# Imperial College London

# Who needs IEEE double precision floating-point?

Eric Kerrigan, Juan Jerez, Stefano Longo, George Constantinides Department of Electrical and Electronic Engineering Department of Aeronautics









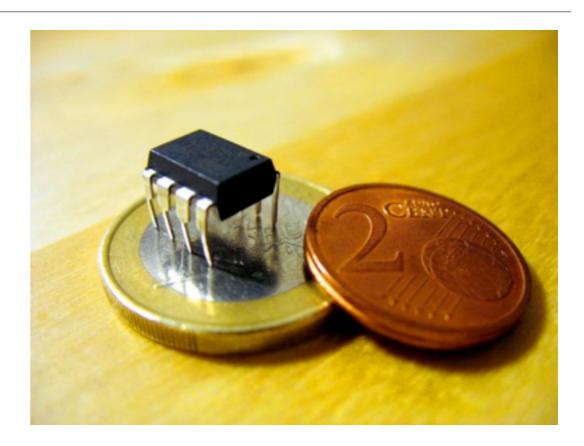


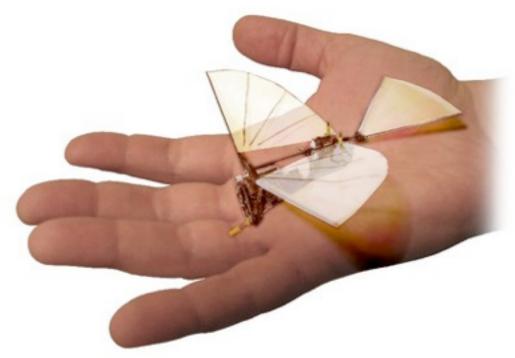


#### We Need New Ideas And We Need It Soon

Can you get a nonlinear model-based predictive controller to do aerobatics of a micro-UAV using a:

- 2 cent, 1 nW microcontroller
- in 5/10/20/50/100 years' time?





# Old Ideas Will Probably Never Work



# Applications for Embedded Optimization (Optimal Control / Estimation / DSP)















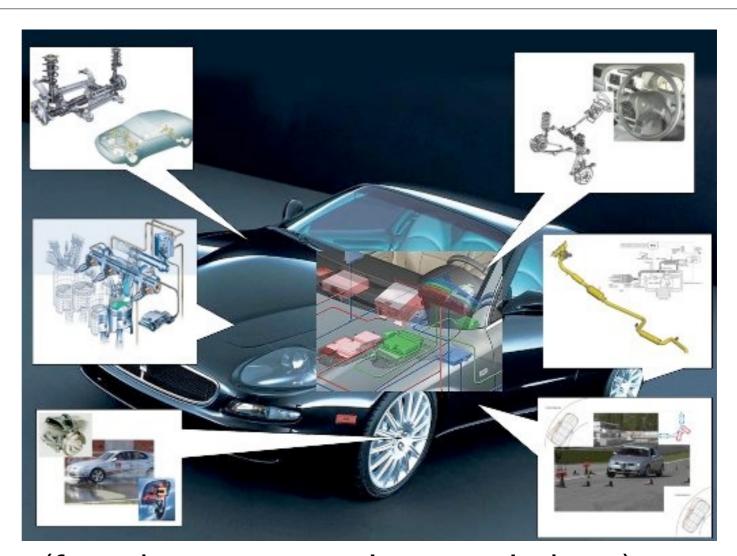






#### Challenges for Embedded Systems

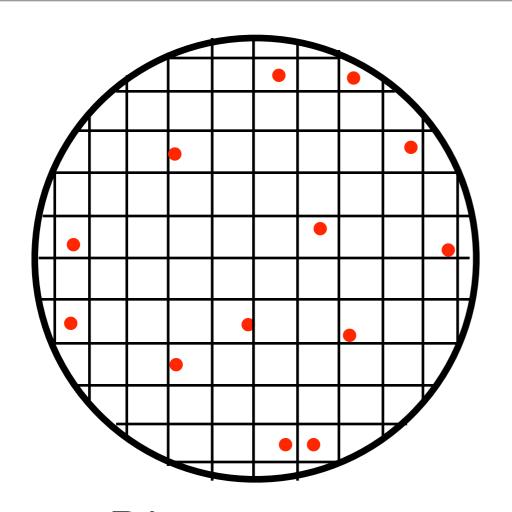
- Cost
- Energy
- Speed
- Reliability



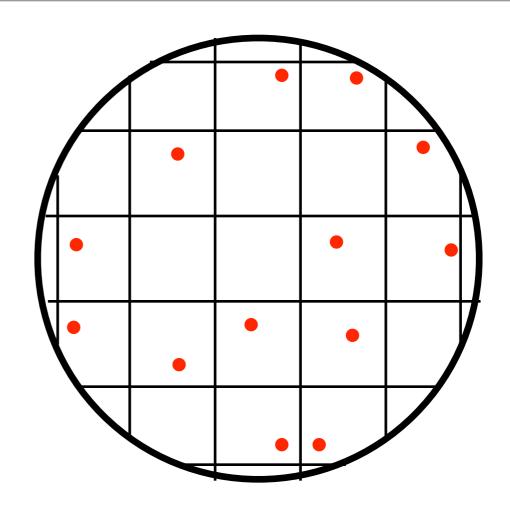
Predictability / real-time (fast is not equal to real-time)

Number representation (e.g. fixed/floating-point, #bits) has a major impact on the design

#### Size is Very Important in Microprocessor Design



Die area = 1 Working = 64



Die area = 4 Working = 4

Cost per die =  $f(area^x)$ ,  $x \in [2,4]$ 

#### Computational Resources for an Adder

#### Xilinx Virtex-7 XT 1140 FPGA:

Number representation	Registers (FFs)	Latency/delay (clock cycles)
double floating-point 52-bit mantissa	1046	14
single floating-point 23-bit mantissa	557	11
fixed-point 53 bits	53	1
fixed-point 24 bits	24	1

Cheap and low power processors often only have fixed-point

# Dynamic Optimization

$$\min_{x(\cdot), u(\cdot), p} J(y(\cdot), x(\cdot), u(\cdot), p)$$

$$F(y(t), \dot{x}(t), x(t), u(t), p, t) = 0, \quad \forall t \in [t_0, t_f)$$

$$G(y(t), \dot{x}(t), x(t), u(t), p, t) \leq 0, \quad \forall t \in [t_0, t_f)$$

Discretized and approximated by finite-dimensional NLP:

$$\min_{\theta} V(\theta)$$

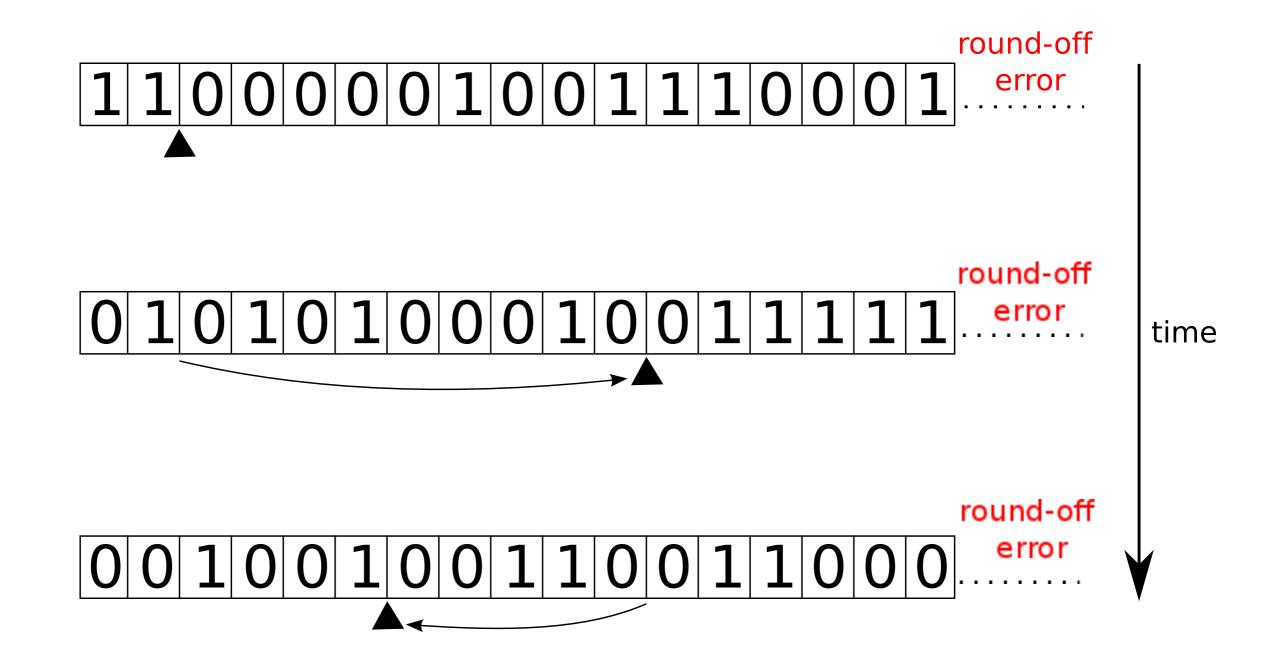
$$f(\theta) = 0$$

$$g(\theta) \le 0$$

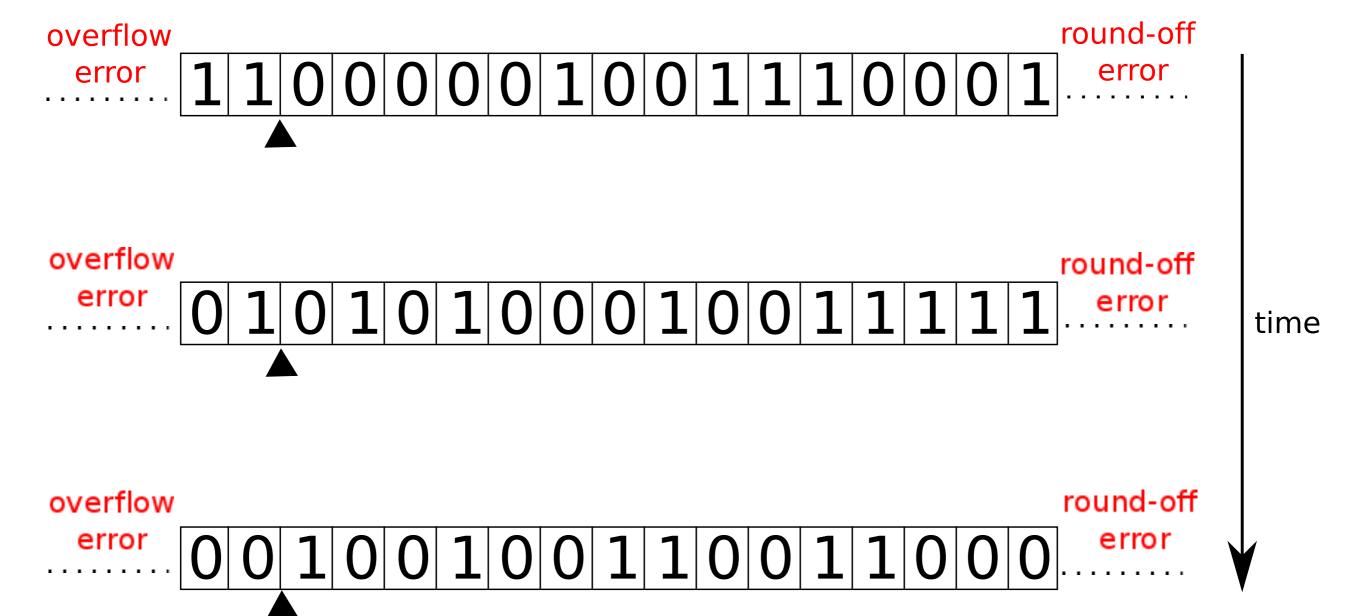
$$\theta \in \mathbb{R}^n, \ f : \mathbb{R}^n \to \mathbb{R}^m, \ g : \mathbb{R}^n \to \mathbb{R}^p$$

Fixed-Point Arithmetic

#### Floating-Point Arithmetic



#### Fixed-Point Arithmetic



#### Challenges for Fixed-Point Arithmetic

- Number of bits for integer and fractional part?
  - Determine worst-case peak values
  - Optimization algorithms are nonlinear and recursive
- Search direction most computationally critical part:

$$\mathsf{A}\xi=\mathsf{b}$$

• Iterative linear solvers preferred: CG, MINRES, GMRES

# Lanczos Algorithm (Kernel of CG/MINRES)

$$\mathbf{Q}_i^T \mathbf{A} \mathbf{Q}_i = \mathbf{T}_i := \begin{bmatrix} \alpha_1 & \beta_1 & & 0 \\ \beta_1 & \alpha_2 & \ddots & \\ & \ddots & \ddots & \beta_{i-1} \\ 0 & & \beta_{i-1} & \alpha_i \end{bmatrix}$$

Given  $q_1$  such that  $||q_1||_2 = 1$  and an initial value  $\beta_0 := 1$  for i = 1 to  $i_{max}$  do

- 1.  $q_i \leftarrow \frac{q_i}{\beta_{i-1}}$
- 2.  $z_i \leftarrow \mathsf{A}q_i$
- 3.  $\alpha_i \leftarrow q_i^T z_i$
- 4.  $q_{i+1} \leftarrow z_i \alpha q_i \beta_{i-1} q_{i-1}$
- 5.  $\beta_i \leftarrow ||q_{i+1}||_2$

#### end for

# Lanczos Algorithm (Kernel of CG/MINRES)

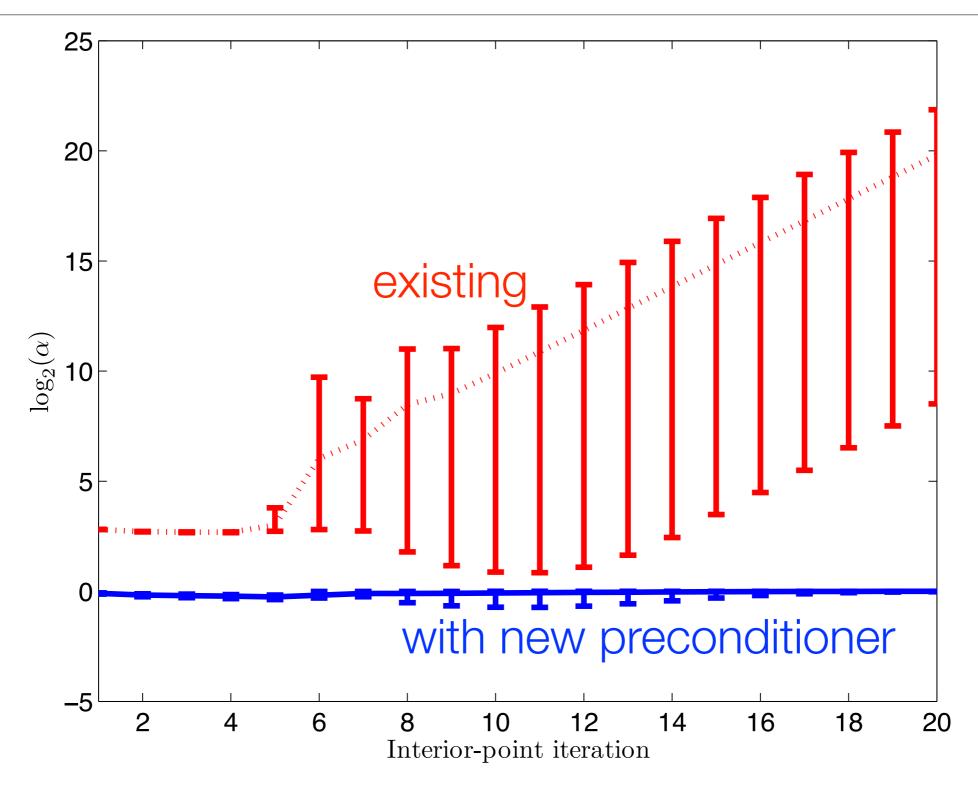
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#### end for

#### Evolution of Variables in Primal-dual Interior Point



#### On-line Diagonal Preconditioner / Scaler

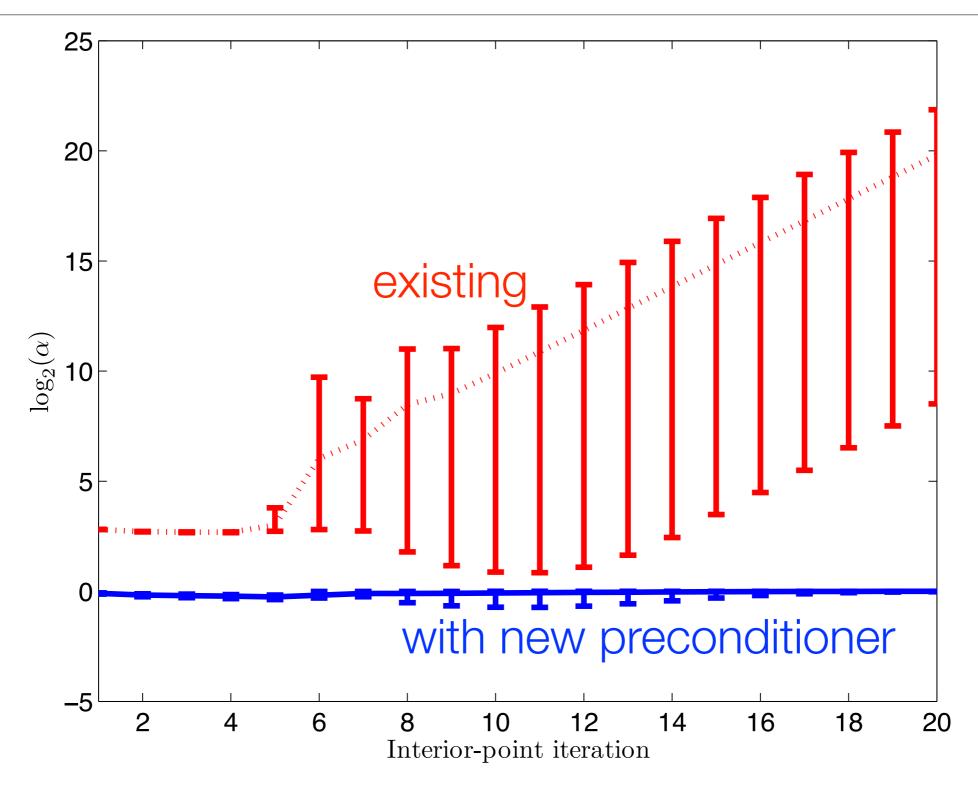
$$\begin{aligned} \mathsf{A}\xi &= \mathsf{b}, \quad \mathsf{A} = \mathsf{A}' \\ \mathsf{S}_{kk} &:= \sum_{j=1}^{\mathsf{N}} |\mathsf{A}_{kj}| \quad \text{(1-norm of row $k$)} \\ \mathsf{S}^{-\frac{1}{2}}\mathsf{A}\mathsf{S}^{-\frac{1}{2}}\psi &= \mathsf{S}^{-\frac{1}{2}}\mathsf{b} \Leftrightarrow \widehat{\mathsf{A}}\psi = \widehat{\mathsf{b}} \Rightarrow \rho\left(\widehat{\mathsf{A}}\right) \leq 1 \\ \xi &= \mathsf{S}^{-\frac{1}{2}}\psi \end{aligned}$$

#### Theorem (Avoiding overflow in fixed-point)

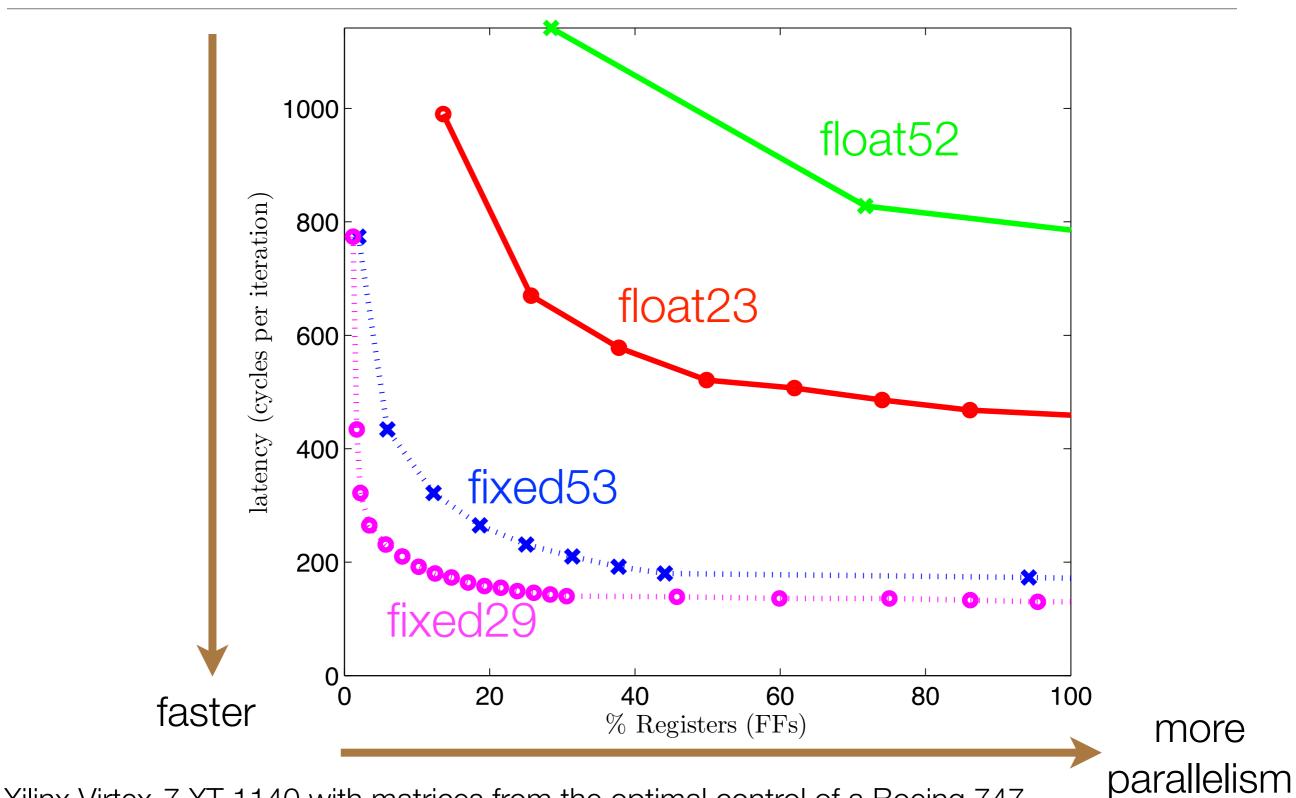
For integer part, need at most 2 bits for all the variables in the Lanczos algorithm and  $\lceil -\log_2(\epsilon) \rceil$  bits for  $1/\beta_i$ 

*Proof*: Proc. IEEE Conference on Decision and Control 2012

#### Evolution of Variables in Primal-dual Interior Point



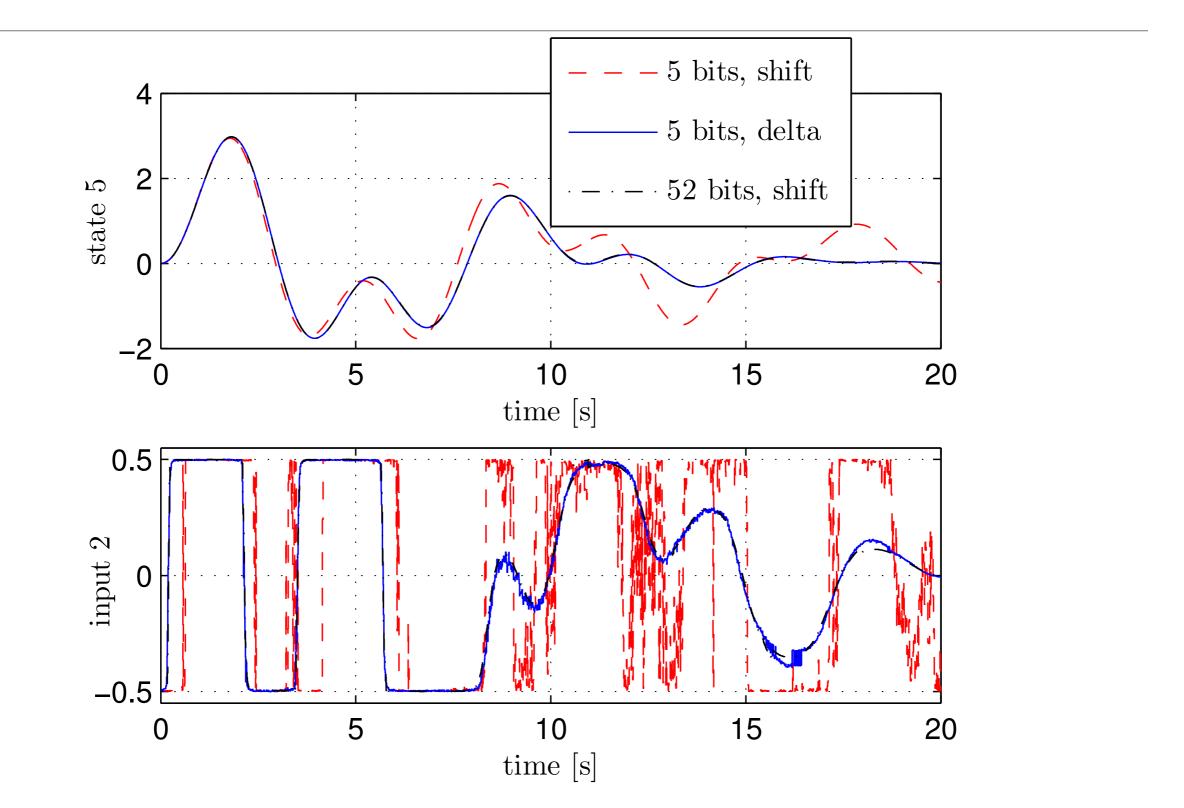
#### Trade-offs on an FPGA (same accuracy)



Xilinx Virtex-7 XT 1140 with matrices from the optimal control of a Boeing 747

Low-Precision Arithmetic

# Optimal Control in Low Precision Arithmetic



Mass-spring system with 3 masses (6 states) and 2 inputs, sample period = 10ms

#### Sampled-data Representation in Shift Form

$$\dot{x}(t) = A_c x(t) + B_c u(t)$$

Sample period h and piecewise constant input (ZOH):

$$u(t) = u(kh) =: u_k, \quad \forall t \in [kh, kh + h)$$

Exact solution/discrete-time model to compute  $x_k := x(kh)$ 

$$x_{k+1} = A_s x_k + B_s u_k$$

$$A_s := e^{A_c h} = I + A_c h + \frac{(A_c h)^2}{2!} + \frac{(A_c h)^3}{3!} + \dots$$

$$\lim_{\|A_c h\| \to 0} A_s = I, \quad \lim_{\|A_c h\| \to 0} B_s = 0$$

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$$u(t) = u(kh) =: u_k, \quad \forall t \in [kh, kh + h)$$

Exact solution/discrete-time model to compute  $x_k := x(kh)$ 

$$((x_{k+1} - x_k)/h = (A_s x_k + B_s u_k - x_k)/h)$$

$$A_s := e^{A_c h} = I + A_c h + \frac{(A_c h)^2}{2!} + \frac{(A_c h)^3}{3!} + \dots$$

$$\lim_{\|A_c h\| \to 0} A_s = I, \quad \lim_{\|A_c h\| \to 0} B_s = 0$$

#### Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$\frac{x_{k+1} - x_k}{h} = \frac{(A_s - I)}{h} x_k + \frac{B_s}{h} u_k$$

#### Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$\frac{x_{k+1} - x_k}{h} = A_\delta x_k + B_\delta u_k$$

#### Sampled-data Representation in Delta Form

Middleton and Goodwin (IEEE TAC, 1986):

$$\frac{x_{k+1} - x_k}{h} = A_{\delta} x_k + B_{\delta} u_k$$

$$A_{\delta} = A_c + \frac{A_c^2 h}{2!} + \frac{A_c^3 h^2}{3!} + \dots$$

$$\lim_{\|A_c h\| \to 0} A_{\delta} = A_c, \quad \lim_{\|A_c h\| \to 0} B_{\delta} = B_c$$

**Equivalent** to shift form in **infinite** precision arithmetic **Different** from shift form in **finite** precision arithmetic

#### Optimization Problem Using Shift Form

$$\min_{\theta} \sum_{k=0}^{N-1} \ell_k(x_k, u_k)$$

$$\theta := \begin{bmatrix} u_0' & x_1' & u_1' & x_2' & \cdots & u_{N-1}' & x_N' \end{bmatrix}'$$

subject to

$$x_0 = \hat{x},$$
  $x_{k+1} = A_s x_k + B_s u_k, \quad \forall k \in \{0, 1, \dots, N-1\}$   $Jx_k + Eu_k \le d, \quad \forall k \in \{0, 1, \dots, N-1\}$ 

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$$\delta_k := \frac{x_{k+1} - x_k}{h} = A_\delta x_k + B_\delta u_k$$

# Optimization Problem Using Delta Form

$$\min_{\theta} \ \sum_{k=0}^{N-1} \ell_k(x_k,u_k)$$
 
$$\theta:= \begin{bmatrix} u_0' & \delta_0' & x_1' & u_1' & \delta_1' & x_2' & \cdots & u_{N-1}' & \delta_{N-1}' & x_N' \end{bmatrix}'$$
 subject to 
$$x_0=\hat{x},$$

$$\begin{cases}
\delta_k = A_{\delta} x_k + B_{\delta} u_k, & \forall k \in \{0, 1, \dots, N - 1\} \\
x_{k+1} = x_k + h \delta_k, & \forall k \in \{0, 1, \dots, N - 1\} \\
J x_k + E u_k \le d, & \forall k \in \{0, 1, \dots, N - 1\}
\end{cases}$$

$$\delta_k := \frac{x_{k+1} - x_k}{h} = A_\delta x_k + B_\delta u_k$$

#### Optimization Problem Using Delta Form

$$\min_{\theta} \sum_{k=0}^{N-1} \ell_k(x_k, u_k)$$
 
$$\theta := \begin{bmatrix} u_0' & \delta_0' & x_1' & u_1' & \delta_1' & x_2' & \cdots & u_{N-1}' & \delta_{N-1}' & x_N' \end{bmatrix}'$$
 subject to 
$$x_0 = \hat{x},$$
 
$$\delta_k = A_\delta x_k + B_\delta u_k, \quad \forall k \in \{0, 1, \dots, N-1\}$$

$$\begin{cases} \delta_k = A_{\delta} x_k + B_{\delta} u_k, & \forall k \in \{0, 1, \dots, N - 1\} \\ x_{k+1} = x_k + h \delta_k, & \forall k \in \{0, 1, \dots, N - 1\} \\ J x_k + E u_k \le d, & \forall k \in \{0, 1, \dots, N - 1\} \end{cases}$$

#### Solving the Optimization Problem

• Solve linearized **KKT system** (Rao, Wright, Rawlings; JOTA, 1998):  $A\xi = b$ 

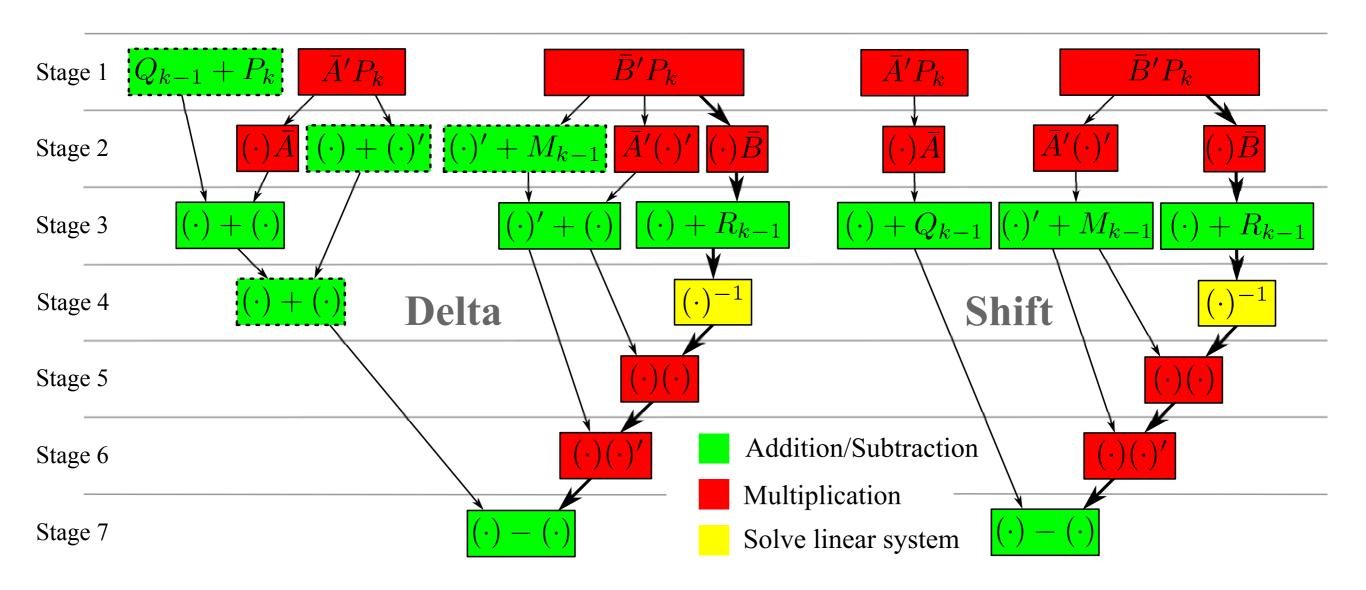
Interleave search direction variables:

$$\xi := \begin{bmatrix} \Delta u_0' & \Delta \gamma_0' & \Delta \delta_0' & \Delta \lambda_1' & \Delta x_1' & \cdots & \Delta x_N' \end{bmatrix}'$$

Block elimination results in Riccati recursions:

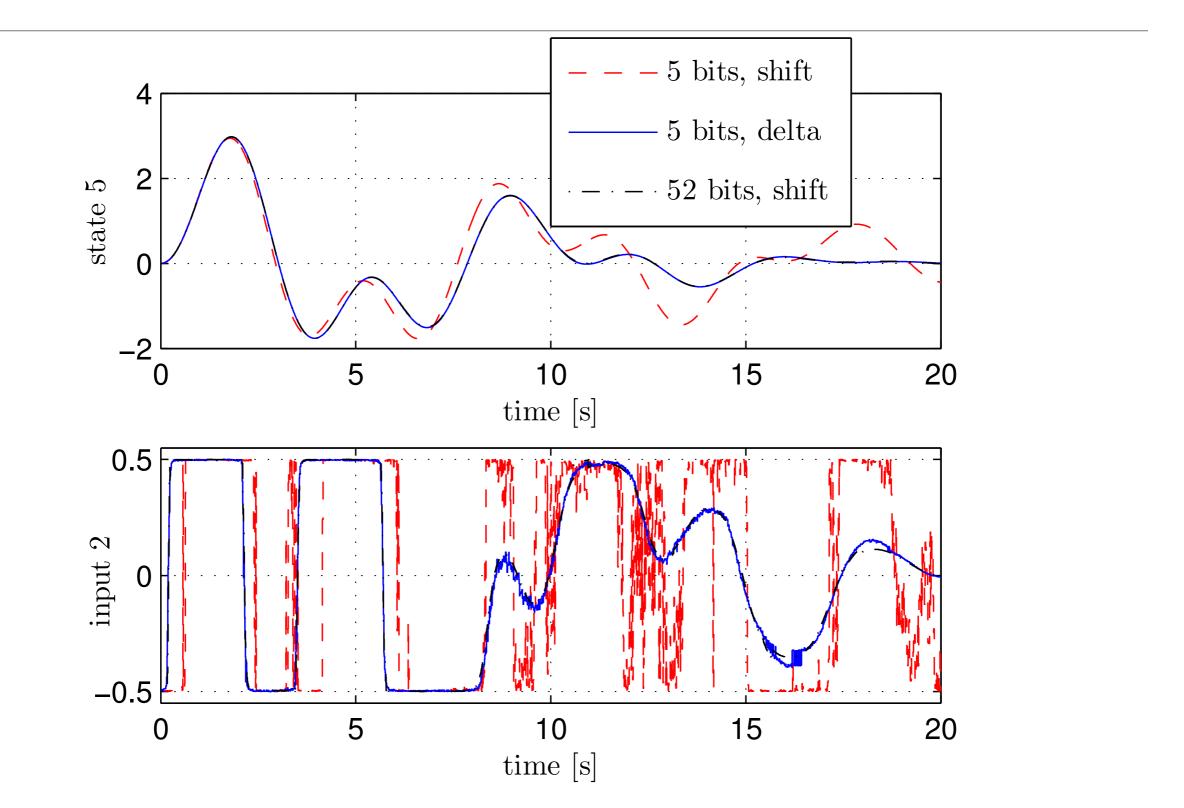
$$P_{k-1} := Q_{k-1} + P_k + h^2 A'_{\delta} P_k A_{\delta} + h A'_{\delta} P_k + h P_k A_{\delta}$$
$$-(M_{k-1} + h^2 A'_{\delta} P_k B_{\delta} + h P_k B_{\delta}) (R_{k-1} + h^2 B'_{\delta} P_k B_{\delta})^{-1}$$
$$(M'_{k-1} + h^2 B'_{\delta} P_k A_{\delta} + h B'_{\delta} P_k)$$

#### Data Dependencies in Riccati Recursion



Same amount of multipliers, adders and computational delay for a custom circuit, e.g. FPGA

# Optimal Control in Low Precision Arithmetic



Mass-spring system with 3 masses (6 states) and 2 inputs, sample period = 10ms

#### Conclusions

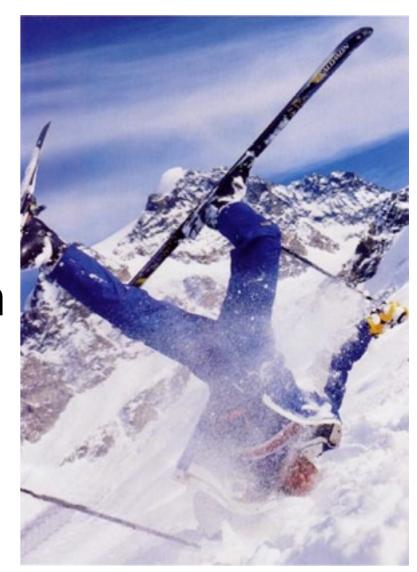
- Number representation major factor that determines cost, energy, computational delay and accuracy
- Fixed-point: Precondition to get tight analytical bounds on variables in Lanczos algorithm to avoid overflow
- Low precision: Sampled-data model and optimization method crucial to successful implementation
- Co-design algorithm and hardware to use "just the right amount" of computational resources

#### Open Research Questions

- Other sampled-data and number representations?
- Nonlinear dynamical models?
- Which algorithms map easily to low precision, fixed-point or other number representations?
- A priori guarantees on accuracy, closed-loop stability, robustness and performance?
- Need control + optimization + numerics + computing

#### Designing for Embedded Systems

"All too often, today's students use laptop [or desktop] computers to perform their computing, which shields them from dealing with any of the physical constraints they will face in the real world. This approach is akin to trying to learn skiing while standing comfortably in the après ski lounge."



Wolf, Cyber-Physical Systems, Computer, 2009.